

The z-Transform

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Outline

1 The z-Transform

Motivation for the z-Transform

The response to a complex exponential z^n is

$$y[n] = H(z)z^n$$

Thus complex exponentials z^n are **eigenfunctions**.

Motivation for the z-Transform

The function $H(z)$

$$H(z) = \sum_{n=-\infty}^{\infty} h[n]z^{-n}$$

is known as the z-transform of $h[n]$.

Don't forget to specify a region of convergence (ROC).

For $z = e^{j\omega}$, $H(z)$ reduces to the Fourier transform of $h[n]$.

An example

The z-transform of the signal $x[n] = a^n u[n]$ is given by

$$X(z) = \frac{1}{1 - az^{-1}}, \quad |z| > |a|$$

Note that for $|a| > 1$, the ROC does not include the unit circle.
The DTFT does not converge.

How about for $x[n] = -a^n u[-n - 1]$?

An example

Another example

$$x[n] = \left[7 \left(\frac{1}{3} \right)^n - 6 \left(\frac{1}{2} \right)^n \right] u[n]$$

Note the ROC is the intersection of both $x_1[n]$ and $x_2[n]$.

$$|z| > \frac{1}{2}$$



Poles and Zeros

$X(z)$ can also be expressed as

$$X(z) = \frac{a_0 \sum_{k=1}^M (1 - c_k z^{-k})}{b_0 \sum_{k=1}^N (1 - d_k z^{-k})}$$

c_k represent the non-zero **zeros** of $X(z)$: $X(z)|_{z=c_k} \rightarrow 0$

d_k represent the non-zero **poles** of $X(z)$: $X(z)|_{z=d_k} \rightarrow \infty$

Note that # zeros = # poles: Don't forget zeros and poles at $z = 0$ or $z = \infty$.



Properties of the ROC

- 1 The ROC of $X(z)$ consists of a **ring** in the z -plane centered about the origin.
- 2 ROC does not contain any poles.
- 3 If $x[n]$ is **left** (right) sided, then ROC is **inward** (outward).
- 4 If $X(z)$ is rational and $x[n]$ is **right-sided**, then ROC is the region in the z -plane **outside the outermost pole**.
Furthermore, if $x[n]$ is causal, the ROC also includes $z = \infty$.



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ROC Example

What are the possible ROCs for the following z-transform?

$$X(z) = \frac{1}{(1 - \frac{1}{3}z^{-1})(1 - 2z^{-1})}$$

The Inverse z-transform

The Inverse z-transform can be obtained via contour integration

$$x[n] = \frac{1}{2\pi j} \oint X(z)z^{n-1} dz$$

There is often an easier way.

The Inverse z-transform

What's the signal $x[n]$ with z-transform

$$X(z) = \frac{3 - \frac{5}{6}z^{-1}}{(1 - \frac{1}{4}z^{-1})(1 - \frac{1}{3}z^{-1})}, \quad |z| > \frac{1}{3}$$

Use **partial fraction expansion**.



The Inverse z-transform

What's the signal $x[n]$ with z-transform

$$X(z) = 4z^2 + 2 + 3z^{-1}, \quad 0 < |z| < \infty$$

Use the **power-series expansion** method.

Properties of the z-transform

Analogous to FT and FS properties.

Convolution

$$x_1[n] * x_2[n] \leftrightarrow X_1(z)X_2(z), \quad \text{ROC} = R_1 \cup R_2$$

Watch how ROC changes.



Characterization of LTI systems using z-transform

- 1 A LTI system is **causal** iff (a) the ROC is the **exterior** of a circle outside the outermost pole, or (b) $H(z)$ is **proper** (degree of numerator does not exceed degree of denominator.)
- 2 A LTI system is **stable** iff the ROC of $H(z)$ includes the **unit circle** $|z| = 1$.
- 3 A **causal** LTI system is **stable** if **all poles** lie inside the unit circle.

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Characterization of LTI systems using z-transform

The z-transform is often in the form

$$\begin{aligned} H(z) &= \frac{Y(z)}{X(z)} \\ &= \frac{\sum_{k=0}^M b_k z^{-k}}{\sum_{k=0}^N a_k z^{-k}} \end{aligned}$$

This comes from the LCCDE

$$\sum_{k=0}^N a_k y[n-k] = \sum_{k=0}^M b_k x[n-k]$$

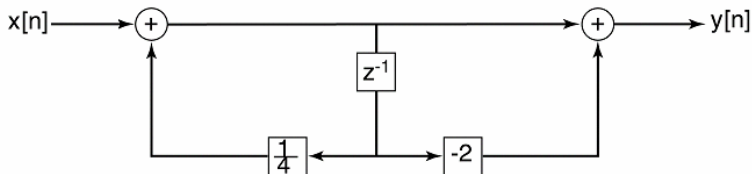
Characterization of LTI systems using z-transform

Find the z-transform $H(z)$ of the an LTI system for which $x[n]$ and $y[n]$ satisfy the LCCDE

$$y[n] - \frac{1}{2}y[n-1] = x[n] + \frac{1}{3}x[n-1]$$

Block diagram representation

Find the z-transform $H(z)$ of the an LTI system for which $x[n]$ is the input and $y[n]$ is the output.



Find the unit sample response $h[n]$.

