Tilted Matching for Feedback Channels

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Feedback for Error Control

- Feedback does not increase the capacity of DMC.
- Feedback is often used to correct errors via architectural changes:

Example: Yamamoto & Ito ' 79



Burnashev'76: optimal error exponent for variable-length codes

Why Variable Length Codes are Nice?

- Ack/NAck/Retransmission is much more efficient than Forward Error Correction without feedback;
- What is hidden from the variable code length?
 - Retransmission costs ignored in average transmission time;
 - Forward transmission remains the same as the no-feedback case;
 - Do not try to recover from a partial error;
- Recent extensions:
 - ► Fixed length block codes with erasures: Nakiboglu & Z'09

 Non-block codes with fixed or variable delays: Draper & Sahai'08

Feedback for Forward Error Correction

- Feedback communication schemes that employ "incrementally" tuned encoding
 - Noise variance reduction in AWGN: [Elias, 56], [Schalkwijk & Kailath, 66],[Schalkwijk, 66, 68]
 - Posterior matching: [Horstein, 63], [Shayevitz & Feder, 07,08]

- Dumped down encoding: [Zigangirov, 70], [D' yachkov, 75]
- Why are these important? Dynamic information exchange.

 Communication as "driving the belief" at the receiver (Coleman)



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- Decision region and reward at the end of the block
- Encoding function does not depend on the correct message, solve parallel problems
- Randomness from the channel
- With feedback (or noisy feedback), encoding can depend on the current belief



Solutions of the Dynamic Coding Problem

- Variety of approximate solutions, depending on how to approximate the reward function, and how to average.
- Example: Posterior matching, Shayevitz & Feder'08

 $X_{t+1} = F_X^{-1}(F(M|y^t))$ is capacity achieving

Coleman'09: This is the optimal solution when using average K-L divergence as reward;

Our question: What is the dynamic problem that leads to the optimal error exponent? What are we ignoring when doing block codes?

Error Exponents for K-ary Symmetric Channels with Feedback



Figure: k-ary symmetric channel with k = 4 and $\epsilon = 3/4$.

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Notations



Encoder:

$$\mathbb{X}_t(m, Y^{t-1}) : \mathcal{M} \times \mathcal{Y}^{t-1} \to \mathcal{X} \qquad t \leq n$$

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Decoder:

$$\hat{m}(Y^n): \qquad \mathcal{Y}^n \to \mathcal{M}$$

• Knowledge at time $t, \varphi_t(\cdot) = \mathbf{P}[m = \cdot | y^t]$

$$P_{\mathbf{e}} = \sum_{y^{n}} \mathbf{P}[y^{n}] \sum_{k} \varphi_{n}(k) \mathbb{I}\{\hat{m}(y^{n}) \neq k\}$$
$$= \mathbf{E}\left[\sum_{k} \varphi_{n}(k) \mathbb{I}\{\hat{m}(Y^{n}) \neq k\}\right]$$

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Rearrange:

$$P_{\mathbf{e}} \leq \mathbf{E} \left[\sum_{k} \varphi_{n}(k)^{1-\rho\eta} (\sum_{l \neq k} \varphi_{n}(l)^{\eta})^{\rho} \right] = \mathbf{E} \left[\zeta \left(\varphi_{n}(\cdot), \eta, \rho \right) \right]$$

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Smoothed Problem

- $\varphi_t(\cdot)$ summarizes the knowledge up to time *t*;
- $\triangleright \zeta_n \triangleq \sum_k \varphi_n(k)^{1-\rho\eta} (\sum_{l \neq k} \varphi_n(l)^{\eta})^{\rho};$
- ζ_t similarly defined;
- $\triangleright \ \zeta_0 = e^{nR\rho};$
- At the end of the block: $P_{\mathbf{e}} \leq \mathbf{E}[\zeta_n];$
- ▶ Design goal: at time *t*, with knowledge $\varphi_t(\cdot)$ and ζ_t , make ζ_{t+1} as small as possible, on average.

Approximations exponentially tight.

Uniform Progress Assumptions

Suppose there exists an encoding scheme such that

$$\mathsf{E}\left[\zeta_{t+1}|\boldsymbol{y}^t\right] \leq \alpha(\rho,\eta)\zeta_t, \qquad \forall t, \boldsymbol{y}^t$$

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Then

$$P_{\mathbf{e}} \leq \mathbf{E}[\zeta_n]$$

$$\leq \mathbf{E}\left[\mathbf{E}\left[\zeta_n | Y^{n-1}\right]\right]$$

$$\leq \alpha(\rho, \eta) \mathbf{E}[\zeta_{n-1}]$$

$$\leq \alpha(\rho, \eta)^n \mathbf{E}[\zeta_0]$$

$$\leq \alpha(\rho, \eta)^n e^{nR\rho}$$

Random Coding Without Feedback

Recall

$$\mathsf{E}\left[\zeta_{t+1}|y^t\right] \leq \alpha(\rho,\eta)\zeta_t, \qquad \forall t, y^t$$

where ζ_{t+1} depends on Y_{t+1} , and X_{t+1} ;

▶ For random coding, and any $\rho \in [0, 1]$, take $\eta = \frac{1}{1+\rho}$,

$$\mathbf{E}\left[\zeta_{t+1}|y^t\right] \le e^{-E_0(P,\rho)}\zeta_t$$

where $E_0(P, \rho) = -\log \sum_{y} (\sum_{x} W(y|x)^{\frac{1}{1+\rho}} P(x))^{1+\rho}$.

Uniform progression is natural when there is no feedback

Tilted Matching for Channel with Feedback

- Potentially with feedback, we can design coding map at each time t to depend on y^t, and to minimize E [ζ_n].
- We focus on coding schemes that only depend on φ_t(·), ρ, and η, and follow the uniform progress assumption.
- The optimal choice, under assumption (*), is a posterior matching of φ_t(·)^η, instead of φ_t(·) itself.
- (*): every message has a posterior probability exponentially small, valid at the early stage

- This achieves the best known error exponent
 - achieves sphere packing exponent for high rates;
 - optimizing over η necessary;

Discussions

- Interpretations of the parameters ρ and η .
 - ρ is similar to Gallager's, to control the union bound, set up threshold between the correct and incorrect messages;
 - η is a slowing down factor, making $\varphi_t(\cdot)$ "flatter".
 - How much do we trust the history at time t;
 - Avoid a single message dominate $\varphi_t(\cdot)$ too early;
- Beyond the uniform progress assumption, dynamic assignment of η.