

BROADCASTING BITS THROUGH QUEUES

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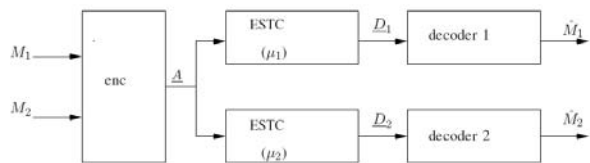
STATUS QUO

- Bits Thru Queues [1]: use **timing** as an unused degree of freedom to communicate.
- [1] Developed closed form expression for capacity of the Exponential Server Timing Channel (ESTC).
- Capacity Region of the ESTC multiple access channel (MAC) found in [2]
- What about the capacity region of a canonical Broadcast Timing Channel?

SET UP

• Encoder **controls** the packet timings that serve as the **arrival process** to the two ESTC queuing systems.

• Inter-arrival times $\{A_i\}_{i=1}^n$ must satisfy a rate constraint: $E\left[\frac{1}{n}\sum_{i=1}^n A_i\right] = \frac{1}{\lambda}$

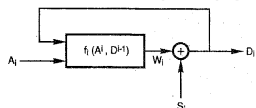


$$\mu_1 > \mu_2 > \lambda$$

• DT model of each first-come, first-serve (FCFS) queue can be represented as[3]

$$D_{1,s} = W_s(\underline{D}_1^{-1}, \underline{A}) + S_{1,s}$$

$$D_{2,s} = W_s(\underline{D}_2^{-1}, \underline{A}) + S_{2,s}$$



Idling Times

Exponential Service Times

$$W_s(\underline{D}, \underline{A}) \triangleq \max\left(0, \sum_{j=1}^i A_j - \sum_{j=1}^{i-1} D_j\right)$$

$$\{S_{1,s}\}_{s=1}^n \sim \prod_{i=1}^n e_{\mu_1}(s_{1,i})$$

$$\{S_{2,s}\}_{s=1}^n \sim \prod_{i=1}^n e_{\mu_2}(s_{2,i})$$

• **Infinite Divisibility** of the exponential distribution [3]: if $E_{b_k} \sim e_{b_k}$ and $b_1 < b_2$, then

$$E_{b_2} = E_{b_1} + M_{b_1 \rightarrow b_2} \quad \text{where} \quad M_{b_1 \rightarrow b_2} \triangleq \begin{cases} 0 & \text{w/ Pr } \frac{b_2}{b_1} \\ \sim e_{b_2}(m) & \text{w/ Pr } 1 - \frac{b_2}{b_1} \end{cases}$$

$$E[M_{b_1 \rightarrow b_2}] = b_2 - b_1 \triangleq a_{b_1 \rightarrow b_2} \quad I(M_{\mu \rightarrow \lambda}; M_{\mu \rightarrow \lambda} + S_\mu) = \log\left(1 + \frac{a_{\mu \rightarrow \lambda}}{b_\mu}\right)$$

$$E[E_k] = b$$

ACHIEVABLE REGION w/ FEEDBACK

• With feedback: encoder can control idling times. Use **rate-splitting** analogous to the Gaussian case [4]

$$U_1 \leftrightarrow \{W^1_1, W^1_2\}$$

$$U_2 \leftrightarrow \{W^2_1, W^2_2\}$$

$$W^1_{1,i} = W^1_{1,i} + W^2_{1,i}$$

$$W^2_{2,i} = W^1_{2,i} + W^2_{2,i}$$

$$b_1 \triangleq \frac{1}{\mu_1}, \quad b_2 \triangleq \frac{1}{\mu_2}$$

$$a_1 \triangleq \frac{1}{\lambda} - b_1, \quad a_2 \triangleq \frac{1}{\lambda} - b_2$$

Decoder dynamics

$$D_{1,i} = W^1_{1,i} + W^2_{1,i} + S_{1,i}$$

$$D_{2,i} = W^1_{2,i} + W^2_{2,i} + S_{2,i}$$

$$E[W^1_{1,i}] = \alpha a_1, \quad E[W^2_{1,i}] = (1 - \alpha)a_1$$

$$E[W^1_{2,i}] = \alpha a_2, \quad E[W^2_{2,i}] = (1 - \alpha)a_2$$

$$= \underbrace{W^2_{2,i}}_{E_{b_2 \rightarrow b_1}} + \underbrace{[W^1_{2,i} + S_{2,i}]}_{E_{b_1}}$$

• Achievable Rate region using rate-splitting and onion-peeling decoding

Decoder 2 treats message $W^1_{2,i}$ as noise:

$$R_2 = I(W^2_2; D_2)$$

$$= I(W^2_2; W^2_2 + [W^1_2 + S_2])$$

$$= \log\left(1 + \frac{(1 - \alpha)a_2}{b_2 + \alpha a_2}\right)$$

Decoder 1 first decodes U_2 then onion-peels W^2_1 out:

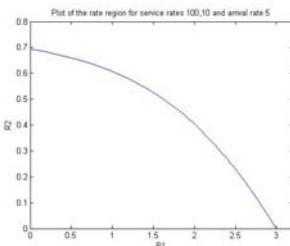
$$R_1 = I(W^1_1; D_1 | W^2_1)$$

$$= I(W^1_1; W^1_1 + W^2_1 + S_1 | W^2_1)$$

$$= I(W^1_1; W^1_1 + S_1)$$

$$= \log\left(1 + \frac{\alpha a_1}{b_1}\right)$$

$$\mathcal{R} = \{R_1(\alpha), R_2(\alpha) : \alpha \in [0, 1]\}$$



CONVERSE IN LOW SNR REGIME

• The structure of the converse is in the spirit of the Bergman's converse [4] for the Gaussian Broadcast channel

Proof by contradiction: Assume (R_1, R_2) in outer exterior of \mathcal{R} .

USER 1

$$(i) \quad h(\underline{D}_1 | U_2) \geq n[1 + \log(b_1 + \alpha a_1)] - n\epsilon_1$$

$$\theta = \frac{1}{\mu}, \quad \theta_0 = \frac{1}{\lambda} \quad \text{Consider } \{\text{ESTC}(\theta)\} \text{ for a fixed } P(\underline{A}).$$

• Queuing Stability Corollary (*): $f_{D^*}(d^*; \theta_0) = \prod_{i=1}^n e_{\lambda_i}(d_i)$

• $\mathcal{H}_{D^*}(\theta) \triangleq \lim_{n \rightarrow \infty} \frac{1}{n} h(f_{D^*}(d^*; \theta))$

• Timing Entropy Power: $g(\theta) \triangleq \exp(\mathcal{H}_{D^*}(\theta) - 1)$

• Maximum-Entropy (rate of \underline{D} is λ): $g(\theta) \leq g(\frac{1}{\lambda}) = \frac{1}{\lambda}$

USER 2

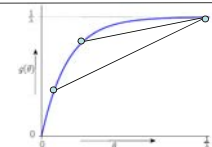
$$(ii) \Rightarrow h(\underline{D}_2) \geq n \log(b_2 + \alpha a_2) - n \log(b_2 + \alpha a_2) + n\delta + h(\underline{D}_2 | U_2) - n\epsilon_2$$

Converse follows directly from the **concavity** of g :

$$\frac{g(\theta_2) - g(\theta_1)}{\theta_2 - \theta_1} \geq \frac{g(\theta_2) - g(\frac{1}{\lambda})}{\theta_2 - \frac{1}{\lambda}} \geq \theta_2 + \alpha_2$$

$$g(\theta_2) U_2 \geq (1 - \alpha)(\theta_2 - \theta_1) + g(\theta_1) U_2 \stackrel{(i)}{\geq} \theta_2 + \alpha_2$$

$$g(\theta_2) \stackrel{(i)}{\geq} \theta_2 + \alpha_2 = \frac{1}{\lambda}$$



Concavity of g in the low-SNR regime:

$$D(\theta) \triangleq \lim_{n \rightarrow \infty} \frac{1}{n} D(f_{D^*}(d^*; \theta) || f_{D^*}(d^*; \theta_0)) \quad \text{Relation between KL divergence and Fisher Information [5]:}$$

$$g'(\theta) = g(\theta) \mathcal{H}_{D^*}(\theta) \Rightarrow \mathcal{H}_{D^*}(\theta) = -D'(\theta) + C \quad \mathcal{H}'_D(\theta_0) = -D''(\theta_0) = 0$$

$$g''(\theta) = g(\theta) [\mathcal{H}'_{D^*}(\theta) + (\mathcal{H}_{D^*}(\theta))^2] (***) \Rightarrow \mathcal{H}'_{D^*}(\theta) = -D''(\theta) \quad \mathcal{H}'_{D^*}(\theta_0) = -D''(\theta_0) = -F\left(\frac{1}{\lambda}\right) \leq 0$$

$$\Rightarrow \mathcal{H}'_{D^*}(\theta) = -D''(\theta) \Rightarrow g''(\theta) \stackrel{(***)}{\leq} 0$$

DISCUSSION

• The structure of the broadcasting bits through ESTC queues problem is shown to be **intimately related** to the **AWGN broadcast** problem [4].

• We have developed an appropriate **entropy power for timing channels** and derived its **concavity** in the low-SNR regime

IMPLICATIONS, EXTENSIONS, AND FUTURE WORK

• Explore the concavity question beyond just the low SNR regime using information geometry.

• Explore achievable schemes without feedback that achieve the same rates

REFERENCES

- [1] V. Anantharam and S. Verdú, "Bits through queues," *IEEE Transactions on Information Theory*, vol. 42, no. 1, pp. 4–18, 1996.
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- [5] S. Kullback. *Information Theory and Statistics*. Wiley, New York, 1959.